



Convolution

L is time-invariant / shift-invariant

$$W(x) = (LV)(x)$$

$$T_h W(x) = (L T_h V)(x)$$

a linear system given by convolution is time-invariant

$$L v = h * v$$

$$(Lv)(x) = \int_{-\infty}^{+\infty} h(x-y) v(y) dy$$

Since $(Lv)(x) = \int_{-\infty}^{+\infty} h(x-y) v(y) dy$, can conclude the impulse-response is $h(x-y)$

Since $h * (T_h v) = T_h (h * v)$ the linear system $w = Lv$ is time-invariant

a time-invariant system is given by convolution

$$(Lv)(x) = \int_{-\infty}^{+\infty} (L \delta_y)(x) v(y) dy$$

$$\text{let } (L \delta_y)(x) = h(x-y)$$

$$\text{then } (L \delta_y)(x) = (T_y (L \delta))(x) = (L \delta)(x-y) = h(x-y)$$

$$(Lv)(x) = \int_{-\infty}^{+\infty} (L \delta_y)(x) v(y) dy = \int_{-\infty}^{+\infty} (L \delta)(x-y) v(y) dy = \int_{-\infty}^{+\infty} h(x-y) v(y) dy$$

the impulse-response of a Linear Time-Invariant (LTI) system $h(x-y) = (L \delta)(x-y)$

Some considerations hold for discrete systems.

$W = LV$ is given by matrix multiplication

$$L \text{ is an LTI iff } w = h * v \quad (h = L \delta \quad h[n-m] = L \delta_m)$$

$W = A \cdot V$ A has a special form for time-invariant systems

eg.

$$h = (1, 2, 3, 4)$$

$$W = h * V \quad (= A \cdot V \quad \text{where } A =)$$

$$W(x) = \sum_y A_{xy} \delta_y(x) \cdot V(y)$$

$$A \text{ has columns } \begin{bmatrix} A \cdot \delta_0 \\ A \cdot \delta_1 \\ A \cdot \delta_2 \\ A \cdot \delta_3 \end{bmatrix}$$

Since the linear system is given by convolution

$$A \cdot \delta_0 = h * \delta_0 = (1, 2, 3, 4)$$

$$A \cdot \delta_1 = h * \delta_1 = (4, 1, 2, 3)$$

$$A \cdot \delta_2 = h * \delta_2 = (3, 4, 1, 2)$$

$$A = \begin{bmatrix} 1 & 4 & 3 & 2 \\ 2 & 1 & 4 & 3 \\ 3 & 2 & 1 & 4 \\ 4 & 3 & 2 & 1 \end{bmatrix} \begin{bmatrix} V \\ \vdots \end{bmatrix} \begin{matrix} h \\ (1 \ 2 \ 3 \ 4) \\ (\cdot \cdot \cdot \cdot) \end{matrix}$$

· LTI system and Fourier Transform

time invariant system $(Lv)(x) = \int_{-\infty}^{+\infty} (L\delta_y)(x) v(y) dy = \int_{-\infty}^{+\infty} (L\delta)(x-y) v(y) dy = L\delta * v$

$$W = h * v$$

impulse-response

$$W(s) = H(s) \cdot V(s)$$

transfer function

$$(f * g) = (fg)(f)(g)$$

for LTI system, in the frequency domain, it's described by direct proportionality

$$W = h * v$$

$$W(s) = H(s) \cdot V(s)$$

input complex exponential $v(x) = e^{i\omega x}$, what's $(Lv)(x)$?

$$f(e^{i\omega x}) = \delta$$

$$\langle f e^{i\omega x}, \phi \rangle = \langle e^{i\omega x}, f\phi \rangle$$

$$= \int_{-\infty}^{+\infty} e^{i\omega x} (f\phi)(x) dx$$

$$= (f\phi)(\omega) = \langle \delta, \phi \rangle$$

$$W(s) = H(s) \cdot \delta(s) = H(\omega) \cdot \delta(\omega)$$

take the inverse F.T.

$$w(x) = f^{-1}(H(\omega) \cdot \delta(\omega)) = H(\omega) \cdot f^{-1}\delta(\omega) = H(\omega) \cdot e^{i\omega x}$$

$$L(e^{i\omega x}) = H(\omega) \cdot e^{i\omega x}$$

complex exponential $e^{i\omega x}$ is an eigenfunction of any LTI system, with eigenvalue $H(\omega)$ the eigenvalue $H(\omega)$ depends on the system

complex exponentials are eigenfunctions, not sin and cos

$$g. L(\cos \omega x) = L(\frac{1}{2}(e^{i\omega x} + e^{-i\omega x}))$$

$$= \frac{1}{2} [L(e^{i\omega x}) + L(e^{-i\omega x})]$$

$$= \frac{1}{2} (H(\omega) \cdot e^{i\omega x} + H(-\omega) \cdot e^{-i\omega x})$$

if h is real, then $\overline{H(\omega)} = H(-\omega)$

$$= \frac{1}{2} (H(\omega) \cdot e^{i\omega x} + \overline{H(\omega)} \cdot e^{-i\omega x})$$

$$= \text{Real}(H(\omega) e^{i\omega x})$$

$$H(\omega) = |H(\omega)| \cdot e^{i\phi}$$

$$= \text{Real}(|H(\omega)| \cdot e^{i(\omega x + \phi)})$$

$$= |H(\omega)| \cdot \cos(\omega x + \phi) \text{ cos with phase shift}$$

Discrete version

$$w = Lv = h * v$$

$$W[m] = H[m] \cdot V[m]$$

$$\text{input } w^k = [1 \ e^{i\omega \frac{k}{N}} \ e^{i\omega \frac{2k}{N}} \ \dots \ e^{i\omega \frac{(N-1)k}{N}}]$$

$$f w^k = N \cdot \delta_k$$

$$W = H \cdot N \cdot \delta_k = H[k] \cdot N \cdot \delta_k$$

$$\begin{bmatrix} | & | & | & \dots & | \\ w^0 & w^1 & & & w^{N-1} \end{bmatrix} \begin{bmatrix} w^0 \\ w^1 \\ \vdots \\ w^{N-1} \end{bmatrix} = \begin{bmatrix} -w^0 \\ -w^1 \\ \vdots \\ -w^{N-1} \end{bmatrix} \begin{bmatrix} w^0 \\ w^1 \\ \vdots \\ w^{N-1} \end{bmatrix} = \delta_k \text{ (or the generalizing)}$$

take the inverse F.T

$$f^* W = \frac{1}{N} N \cdot \text{HDJ} \cdot W^k$$

$$L(W^k) = H[k] \cdot W^k$$

$$\frac{1}{N} \begin{bmatrix} -W^0 \\ -W^1 \\ \vdots \\ -W^{N-1} \end{bmatrix} \begin{bmatrix} N \cdot b_0 \\ \vdots \\ N \cdot b_{N-1} \end{bmatrix} = \frac{1}{N} \begin{bmatrix} W^0 & W^1 & \dots & W^{N-1} \end{bmatrix} \begin{bmatrix} N \cdot b_0 \\ \vdots \\ N \cdot b_{N-1} \end{bmatrix} = W^k$$

1 $W^0 \dots W^{N-1}$ form an orthogonal basis of eigenvectors for any LTI system

eg $W = h * V \quad h = (1, 2, 3, 4)$

$$W = h * V = \begin{bmatrix} 1 & 4 & 3 & 2 \\ 2 & 1 & 4 & 3 \\ 3 & 2 & 1 & 4 \\ 4 & 3 & 2 & 1 \end{bmatrix} \begin{bmatrix} V \\ \vdots \\ V \end{bmatrix}$$

eigenvectors of the system are eigenvectors of h

$$A = \begin{bmatrix} W^0 & W^1 & W^2 & W^3 \\ | & | & | & | \\ | & | & | & | \end{bmatrix} \begin{bmatrix} H[0] \\ H[1] \\ H[2] \\ H[3] \end{bmatrix} \begin{bmatrix} -W^0 \\ -W^1 \\ -W^2 \\ -W^3 \end{bmatrix}$$

$$H = \sum_k h[k] W^k = \sum_k (h[k]) W^k = \begin{bmatrix} 10 \\ -2+j2 \\ 2 \\ -2-j1 \end{bmatrix}$$